Four-terminal resistances in mesoscopic networks of metallic wires: Weak localisation and correlations

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HIGHLIGHTS

- We consider the electronic transport in multi-terminal mesoscopic networks of weakly disordered metallic wires.
- Weak localisation (WL) to conductance and four-terminal resistances, as well as their correlations, involve integration of the Cooperon/Diffuson over the wires with proper weights.
- Nonlocality of quantum transport can lead to large WL correction to the conductance and large four-terminal conductance fluctuations.

ABSTRACT

We consider the electronic transport in multi-terminal mesoscopic networks of weakly disordered metallic wires. After a brief description of the classical transport, we analyse the weak localisation (WL) correction to the four-terminal resistances, which involves an integration of the Cooperon over the wires with proper weights. We provide an interpretation of these weights in terms of classical transport properties. We illustrate the formalism on examples and show that weak localisation to four-terminal conductances may become large in some situations. In a second part, we study the correlations of four-terminal resistances and show that integration of Diffuson and Cooperon inside the network involves the same weights as the WL. The formulae are applied to multiconnected wire geometries.

1. Introduction

Classical laws of transport on electrical networks have been established by Gustav Kirchhoff in 1845. They rely on three fundamental hypotheses. Two of them are energy and charge conservation leading respectively to the so-called voltage and current laws. The third one is Ohm’s law which states that the current $I_{\mu\nu}$ along a wire $(\mu\nu)$ of the network is proportional to the voltage drop between the vertices $\mu$ and $\nu$ connected by this wire:

$$I_{\mu\nu} = \sigma_{\mu\nu} (V_\nu - V_\mu),$$

with $l_{\mu\nu}$ being the length of the wire, $s$ its cross section and $\sigma_{\mu\nu}$ the Drude conductivity. We now understand that this third hypothesis relies on the assumption of diffusive motion of the charge carriers at the microscopic level, within a classical description. In particular, it is not appropriate to describe quantum effects (like Aharonov–Bohm effect) or non-diffusive regime (ballistic regime, quantum Hall effect). After the fundamental breakthrough proposed by Rolf Landauer to describe the electrical conductance as a transmission coefficient, a generalisation of laws of transport beyond classical transport was highly desirable. It has been formalised in a beautiful work by Markus Büttiker (for reviews, see Refs. [21,27,41,50]).

If we restrict ourselves to the regime of linear transport, a convenient description is to start by introducing the conductance matrix $G$ which relates the currents at the contacts (also called terminals) of the circuit to the values of the voltage at these terminals:

$$I_\rho = \sum_\nu G_{\rho\nu} V_\nu.$$
expressing the conservation of current, and the invariance of the current distribution against a global shift of the potentials (gauges invariance). As a generalisation of Landauer’s formula, elements of the conductance matrix are related to transmission coefficients

$$ G_{\alpha\beta} = \sum_{\mu} C_{\alpha\mu} C_{\beta\mu} = 0, $$

where the factor 2 stands for the spin degeneracy. At this level, the formalism is completely general and no hypothesis is made on the nature of electrical transport which is totally encoded in the transmission coefficients $T_{\alpha\beta}$. These coefficients are related to the scattering matrix which may be determined explicitly within specific models. For example, in the regime of the integer quantum Hall effect, the current is carried by edge states which makes the problem effectively one-dimensional and allows for a simple construction of the scattering matrix [17,20]. More generally, the scattering matrix may be constructed efficiently by assuming strictly one-dimensional character, like in a network of strictly one-dimensional (1D) wires [60]. Another powerful approach applies to devices in which the electron dynamics can be considered as ergodic, leading to a random matrix formulation of quantum scattering [12]. In the present article, we consider the case of metallic samples made of weakly disordered wires such that the electron dynamics is diffusive, as it is the case in narrow metallic wires deposited on a substrate [73] (see also the recent papers of Büttiker [22,14,15,19] (see also [27] and Chapter 5 of [41]).

Büttiker emphasised the importance of the measurement process when a quantum circuit is connected to the outside macroscopic world [16]. The measured resistance is not only a property of the system itself but also depends on the way it is connected to the outside world. Moreover, although the concept of conductance is natural from a theoretical point of view, experiments most frequently deal with voltage measurements: current is injected and collected at two specific contacts and voltages are measured at any pair on contacts playing the role of voltage probes (Fig. 1). Therefore, the relevant quantities characterizing the response of the device are the four-terminal resistances defined as

$$ R_{\alpha\beta} = \frac{V_{\alpha} - V_{\beta}}{I}, $$

with

$$ L_{\lambda} = \begin{cases} I_{\alpha} = I \\ I_{j} = -I \\ I_{j} = 0 & \forall \lambda \neq \alpha, \beta \end{cases} $$

By appropriate inversion of relation (2), Büttiker could relate the four-terminal resistances to the elements of the conductance matrix, therefore to the transmission coefficients [14,19]:

$$ R_{\alpha\beta} = \frac{h T_{\alpha\beta} T_{\alpha\beta} - T_{\alpha\beta} T_{\alpha\beta}}{2 e^2}, $$

where $D$ is any minor of the dimensionless conductance matrix. The expression is valid when all indices are different.

In this paper, we are interested in transport properties of mesoscopic diffusive wires, where quantum interferences lead to small deviations to Ohm’s law. This is the so-called weak-localisation regime. For classical transport, simple application of Ohm’s law leads to the expression of the transmission coefficients, which amounts to classical addition of resistances and conductances (Kirchoff). They are expressed in terms of the elements of a matrix which encodes the conductances of all the links of the network, defined below in Eq. (11).

For a single wire of length $L$, it is well-known that the weak-localisation correction to the classical transmission coefficient can be written as [2]

$$ \Delta T = -\frac{2}{L^2} \int_{0}^{L} dx \, P(x, x), $$

where $\Delta T = \langle T \rangle - T_{\text{class}}$ is an average over disorder configurations. The so-called Cooperon $P(x, x)$ measures the contribution of interfering closed electronic diffusive trajectories. We have shown that in a network of diffusive wires, this simple relation generalises to [61]

$$ \Delta T_{\alpha\beta} = \frac{2}{\xi_{\text{loc}}} \sum_{i} \frac{\partial T_{\text{class}}}{\partial l_{i}} \int_{\text{wire } i} dx \, P(x, x), $$

where $i$ labels all the wires of the network, $\xi_{\text{loc}} = \alpha N_c \xi_c$, is the localisation length in the infinitely long wire with $N_c$ conducting channels [12], $\xi_c$ the elastic mean free path and $\alpha$ a dimensionless parameter of order unity, which will be given below.

From the knowledge of the quantum corrections (8), we will show in this paper that the quantum correction to the classical four-terminal resistance is

$$ \Delta R_{\alpha\beta} = \frac{2}{\xi_{\text{loc}}} \sum_{i} \frac{\partial R_{\text{class}}}{\partial l_{i}} \int_{\text{wire } i} dx \, P(x, x). $$

This expression is quite simple since the weights attached to each wire have a simple interpretation: they express the sensitivity of the classical four-terminal resistance when the resistance of this wire is modified.

Similarly we have found convenient expressions for the correlation functions of the transmission coefficients, from which one can obtain the correlation functions of the four-terminal conductances. Like the weak-localisation correction, these expressions involve contribution of all the wires, which are weighted by similar factors $\frac{\partial R_{\text{class}}}{\partial l_{i}}$, Eqs. (48)–(51). These equations, which are, with Eq. (9), the main results of the article, will be illustrated by several examples in simple devices.

Before going specifically to the analysis of quantum transport in networks of quasi one-dimensional weakly disordered wires, we close the section with some general remarks as the concept of four-terminal resistance (4TR), of which Büttiker has been one of the main promotors, has been extremely fruitful in mesoscopic physics. Let us mention few directions:

- In the early developments of the Landauer–Büttiker approach, the concept of four-terminal resistance has helped clarifying the question of contact resistance. The concept of contact resistance is the “mesoscopic version” of the electric resistance appearing when the electronic fluid is injected from a macroscopic conductor into a small hole (known as Sharvin resistance [55]). The role of contact resistances has been nicely explained in several papers of Büttiker [22,14,15,19] (see also [27] and Chapter 5 of [41]).

- A fundamental aspect of quantum transport concerns the symmetry of transport coefficients: symmetry with respect to

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1 Note that the perturbative approach is valid for $\min(l_{\alpha}, l_{\beta}) \ll \xi_{\text{loc}}$. 

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Fig. 1. The four-terminal resistance $R_{\alpha\beta}$ is the ratio of the voltage between two contacts $\mu$ and $v$ and the current injected at contact $\alpha$ and collected at contact $\beta$. 

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current sources and voltage probes exchange, and symmetry with respect to the reversal of the magnetic field $\mathbf{B}$. Extending these ideas to coherent conductors, Büttiker has demonstrated the symmetry relation for the 4TRs [14,19]:

$$R_{\mu\nu\alpha\beta}(-\mathbf{B}) = R_{\mu\nu\beta\alpha}(\mathbf{B}).$$

(10)

- The concept of 4TR provides an illuminating description of the integer quantum Hall effect [17] from the edge state picture introduced by Halperin [36], as it allows us to compute straightforwardly the longitudinal and Hall resistances (see also Büttiker’s beautiful review article [20]). Furthermore, this framework permits us to analyse in simple terms other more subtle effects, like the scattering between edge states at opposite boundaries due to a constriction, leading to the so-called anomalous Hall effect [45,71,20], or by impurities [18], the description of transport in a Hall cross [34], etc.

- Another issue which has been put forward by Büttiker with others, and which will be central in the present article, concerns the nonlocal nature of quantum transport, and the influence of voltage probes on the transport properties of a coherent conductor. Motivated by a set of experiments in multiconnected metallic wires [13,56] (see also Whasburn and Webb’s review [73]), various authors have analysed the role of voltage probes in devices made of disordered wires by various approaches: Maekawa, Isawa and Ebisawa [49], Büttiker [16], Divincenzo, Kane and Lee [43,28], Chandrasekher, Prober and Santhanam [52,53,25], and Hershfield and Ambegaokar [38,39]. A more general discussion of nonlocality of weak localisation in networks of metallic wires was made possible within the theory developed by us in Ref. [61]. Our formalism has allowed us to study how Altshuler–Aronov–Spivak oscillations of the magnetoconductance are affected by the network geometry [30–32,63,51,54,58] (this was beautifully demonstrated by earlier experiments in arrays of lithium rings by Bishop et al. [29]), or the role of electronic interactions [48,68,58,64,33,59,67,68,23].

The outline of the paper is as follows: in the next section we introduce a classical description of electronic transport in networks of metallic wires. Section 3 is devoted to the analysis of the weak localisation correction, as a warm up exercise preparing the more complicated study of four-terminal resistance correlations presented in Section 4. The main formulae, Eqs. (9), etc., are illustrated on simple cases. Section 5 closes the paper with some concluding remarks.

2. Classical transport in networks

2.1. Electrostatic potential

A typical network is represented in Fig. 2. In this section, we introduce a specific notation in order to distinguish between internal vertices and vertices corresponding to reservoirs, by labelling these latter with a prime.

Before going to the discussion of quantum transport, let us analyse the classical transport in the network. For this purpose we start by solving the Poisson equation $\Delta V(\mathbf{r}) = 0$ for the electrostatic potential $V(\mathbf{r})$ inside the network. Boundary conditions are $V(\mathbf{r}) = V_\alpha$ for $\mathbf{r} \in \text{reservoir } \alpha$. Introducing the coordinate $x$ measuring the distance along the wire $\mu$ (from $\mu$ to $\nu$), the potential varies linearly as $V(\mathbf{r}) = V_\alpha (1-x/l_\mu) + V_\nu x/l_\nu$. The (classical) current density is given by Fick’s law $\mathbf{j}(\mathbf{r}) = -eD\nabla \delta n(\mathbf{r})$, where $D$ is the diffusion constant. The density in excess $\delta n$ is related to the potential through the effective (screened) Coulomb interaction $eV(\mathbf{r}) = (1/\rho_0)\delta n(\mathbf{r})$, leading to $\mathbf{j}(\mathbf{r}) = -\rho_0 \nabla V(\mathbf{r})$, where $\rho_0 = e^2/\epsilon D$ is the Drude conductivity and $\epsilon_0$ the density of states at Fermi energy. The current in the wire $\mu$ is given by (1), hence current conservation at each “internal” vertex $\mu$, $\sum_{\nu \text{ neighbour of } \mu} l_\mu = 0$, may be rewritten as $\sum_\nu (M_0)_{\mu\nu} V_\nu = 0$, where the matrix is defined as

$$(M_0)_{\mu\nu} = \delta_\mu + \sum_\nu a_{\mu\nu} - a_{\nu\mu},$$

(11)

where $a_{\mu\nu}$ is the adjacency matrix element, equals to 1 if a wire connects the two vertices and 0 otherwise; thus it constraints the sum in (11) to run over vertices neighbours of $\mu$. The parameters $\lambda_{\mu\nu}$ have been introduced for convenience for the following and describe connection to reservoirs ($\lambda_{i\mu} = 0$ for any internal vertex and $\lambda_{i\mu} \rightarrow \infty$ for a vertex in a reservoir, cf. Fig. 2). Up to a factor $\rho_0$, the matrix $M_0$ thus simply gathers all the wire conductances $a_{\mu\nu}/l_{\mu
u}$, where $s$ is the cross section of the wires. If we split the vector gathering the electrostatic potentials at the vertices into two parts related to internal and external vertices, $(V_{\text{in}} V_{\text{res}}) = (\cdots, V_\alpha, \cdots, V_\beta, \cdots)$, we may write $(M_0)_{\text{in\text{res}}} V_{\text{in}} = - (M_0)_{\text{in\text{res}}} V_{\text{res}}$, i.e. $V_{\text{res}} = \sum_{\nu \text{ res}} \lambda_{\nu\mu} (M_0^{-1})_{\mu\nu} V_{\text{in}}/l_{\mu
u}$ where the sum runs over the reservoirs. It will be convenient for the following to introduce the notation $P_{\beta\alpha}(x, \beta') = \lambda_{\nu\mu} (M_0^{-1})_{\mu\nu}$, which represents the so-called Diffuson, measured at the two vertices:

$$V_{\alpha} = \sum_{\nu \text{ res}} P_{\beta\alpha}(x, \beta') V_{\beta'}.$$

(12)

The Diffuson is solution of the diffusion equation:

$$-\delta_\mu^2 \delta(x - x') = \delta(x - x')$$

(13)

with Dirichlet boundary conditions at the reservoirs: $P_{\beta}(x, \alpha') = P_{\beta}(x, \alpha) = 0$ for all reservoirs $\alpha'$ (for details, see Appendix of Ref. [59]). Using the linearity of the Diffuson on the wires and that it vanishes at the reservoirs, we write $P_{\beta}(\bullet, \alpha) = (x/l_{\beta}) P_{\beta}(\bullet, \alpha)$ for $x \in \beta$ (the reservoir is at $x = 0$). Hence, we can rewrite the coefficients in (12) as $P_{\beta}(x, \beta')/l_{\beta'} = \delta_\beta P_{\beta}(x, \alpha)$, where $x$ is any position in the wire $\beta$. In the following we will prefer to write the slope of the Diffuson on the wire $\beta'$ as $P_{\beta}(\alpha, \beta')/l_{\beta'} = \delta_\beta P_{\beta}(\alpha, x)$, where $x$ is any position in the wire $\beta$. In the following we will prefer to write the slope of the Diffuson on the wire $\beta'$ as $P_{\beta}(\alpha, \beta')/l_{\beta'} = \delta_\beta P_{\beta}(\alpha, x)$, where $x$ is any position in the wire $\beta'$. In the following we will prefer to write the slope of the Diffuson on the wire $\beta'$ as $P_{\beta}(\alpha, \beta')/l_{\beta'} = \delta_\beta P_{\beta}(\alpha, \beta')/l_{\beta'} = \delta_\beta P_{\beta}(\alpha, x)$, where $x$ is any position in the wire $\beta'$. In the following we will prefer to write the slope of the Diffuson on the wire $\beta'$ as $P_{\beta}(\alpha, \beta')/l_{\beta'} = \delta_\beta P_{\beta}(\alpha, x)$, where $x$ is any position in the wire $\beta'$.
coefficients [61]. The relation with this reference’s notations will be clear by using below this expression for the Diffuson’s slope. Since the vertex $\alpha$ may be any point of the network, we may rewrite (12) as
\[
V(x) = \sum_{\text{res. } \beta} \frac{P_d(x, \beta)}{\ell} V'_{\beta},
\] (14)
the expression (14) is only valid when $x$ is at distance larger than $\ell_c$ from the reservoirs.

2.2. Current distribution and generalised conductances

Integrating the current density $J_\mu(x) = -\sigma_\beta \phi V(x)$ across the section $s$ of the wire $\mu$, we obtain the current under the form
\[
I_\mu = \sum_{\text{res. } \beta} G_{\mu, \beta} V'_{\beta},
\] (15)
where we have introduced
\[
G_{\mu, \beta} = \frac{\left( M_{\mu} \right)_{\beta\mu} - \left( M_0 \right)_{\beta\mu}}{\Delta_{\mu \beta} l_{\mu \beta}}\] for $x \in \mu$.

The quantity $G_{\mu, \beta}$ is a generalised conductance matrix relating the external potentials to the internal currents. They obviously satisfy the symmetry property $G_{\mu, \beta} = G_{\beta, \mu}$. Although the physical interpretation was not provided in Ref. [61], the explicit expression of (16) in terms of the matrix $M_0$ was given:
\[
G_{\mu, \beta} = \left( M_0 \right)_{\beta\mu} - \left( M_0 \right)_{\mu\beta} l_{\mu \beta} / \Delta_{\mu \beta} l_{\mu \beta},
\] (17)
If we consider the case of a wire connected to a reservoir, $\mu \rightarrow \alpha'$, $\mu \rightarrow \alpha$, Eq. (15) coincides with the usual relation between currents and voltage in the terminals:
\[
I_\alpha = I_{\alpha'} = \sum_{\text{res. } \beta} G_{\alpha', \beta} V'_{\beta}.
\] (18)
The conductance matrix is obviously related to the generalised conductances by
\[
G_{\mu, \beta} = G_{\alpha, \beta}',
\] (19)
The expression of the classical conductance matrix in terms of the matrix $M_0$ may be deduced by setting $\mu \rightarrow \alpha'$ in (17); we recover the expression of Ref. [61]:
\[
G_{\mu, \beta}' = -\frac{2e^2}{h} \sigma_N \frac{P_d(x, \beta)}{\ell} = -\frac{2e^2}{h} \sigma_N \frac{M_0}{\ell},
\] (20)
We have used $\sigma_\beta = (2e^2/h) \sigma_N \ell / \ell_c$, where $\ell_c$ is the spin degeneracy, $N_e$ the number of conducting channels and $\sigma_N = V_d / V_{d-1}$ involves the volume of the $d$-dimensional sphere of unit radius (thus $\alpha_1 = 2$, $\alpha_2 = \pi/2$ and $\alpha_3 = 4/3$).

**Example.** As a simple illustration of Eq. (20), we consider the ring of Fig. 3. From the definition (11), we write the internal part of the matrix $M_0$ (i.e. the block related to vertices 1 and 2):

\[
\left( M_0 \right)_{\text{in,in}} = \begin{pmatrix}
1/1_l + 1/1_{c\ell d} & -1/1_{c\ell d} \\
-1/1_{c\ell d} & 1/1_{c\ell d} + 1/1_b
\end{pmatrix}
\] (21)
where $1/1_{c\ell d} = 1/l_1 + 1/l_{c\ell d}$. Eq. (20) leads to
\[
\frac{1}{l_{\beta_{\text{b}}}} \left( M_0 \right)_{12} = \frac{1}{l_u + l_{c\ell d} + l_b}
\] (22)
giving the expected conductance.

2.3. Four-terminal resistances

Using (6), the 4TRs can be deduced from the conductance matrix (20) (see also the discussion in Appendix A). For simple enough networks, the determination of the resistances is however more simple than that of the conductance matrix and does not require the knowledge of the latter, as the simple example analysed just before has shown.

3. Weak localisation

Weak localisation is a small quantum correction to transport coefficients originating from quantum interference between time reversed electronic trajectories [8,2]. The main interest in this small quantum correction to transport coefficients is that it gives a measure of the phase coherence length $l_\varphi$, the fundamental characteristic length scale which sets the boundary between quantum and classical physics. It is worth stressing that there is no intrinsic definition of $l_\varphi$, which can only be obtained by extracting a characteristic length scale from the analysis of a physical quantity sensitive to quantum interference, such as weak localisation. A precise experimental determination of $l_\varphi$ thus requires a perfect knowledge of the functional form of the transport coefficients as a function of the various length scales, the magnetic field, etc.

3.1. Conductance matrix

The weak localisation correction to the conductance matrix elements $G_{\mu, \beta} = -(2e^2/h) l_{\mu \beta}$ is given by [61]
\[
\Delta G_{\mu, \beta} = 2l_{\ell_c} \int_{\text{Network}} dx \phi(x) \partial \partial \phi(x) P_l(x, \beta) - P_l(x, \beta) \partial \partial \phi(x),
\] (23)
where $P_l(x, \beta)$ is the Cooperon, solution of
\[
\left( 1/l_0^2 + D_{\beta}^2 \right) P_l(x, \beta') = -\delta(x - x'),
\] (24)
with $D_{\beta} = \partial_{\beta} - 2i \epsilon \hat{\alpha}(x)$ being the covariant derivative and $l_0$ the phase coherence length. Few remarks:

- The effect of the magnetic field is twofold: (i) in the presence of loops in the network, the Cooperon depends on the magnetic fluxes, which leads to Al'tshuler–Aronov–Spivak oscillations [4,5,10] (see also [63,59]). (ii) The penetration of the magnetic field in a narrow wire of width $w$ can be accounted for through the substitution $1/l_0^2 \rightarrow 1/l_0^2 + 1/L_0^2$, where the magnetic length is $L_0 = [\sqrt{3}/2\pi] \phi_0 (\phi_0 = w$ the quantum flux).
- The expression (23) is of great generality: it is not only valid for a system made of quasi-1D wires (network) but also assume that the contacts have a quasi-1D geometry. In such a more general situation, the derivatives should simply be replaced by gradients $\partial \partial / \phi \phi / \hat{\theta} P_l(y, \beta) \rightarrow \nabla P_l(y, \beta) \nabla P_l(y, \beta)$.
- Expression (23) is valid for $\alpha' \neq \beta'$. The direct diagrammatic calculation of the reflection probability is more difficult and involves a description of the matching between the metallic system and the contacts which goes beyond the derivation of
However the WL correction to the diagonal conductance matrix elements can always be deduced by using current conservation \( \sum \Delta G_{\mu\nu} = 0 \).

- Interestingly, we see that the change of each wire is weighted by the “internal conductances” introduced above:

\[
\Delta T_{\mu\nu} = \frac{2}{(2\pi \hbar)^2} \sum_{\mu \neq \nu} \frac{\Delta G_{\mu\nu}^{class}}{\partial \phi_{\mu\nu}} \int_{\text{wire}_{\mu \neq \nu}} dx \, P(x, x).
\]

This expression shows that a uniform integration of the Cooperon in the network is possible only if the distribution of the classical currents in the wires is uniform.

More conveniently, we showed in Ref. [61] that these weights are related to the derivatives of the classical conductance matrix, Eq. (8), or equivalently

\[
\Delta G_{\mu\nu}^{\text{class}} = \frac{2}{(2\pi \hbar)^2} \sum_{\mu \neq \nu} \frac{\Delta G_{\mu\nu}^{\text{class}}}{\partial \phi_{\mu\nu}} \int_{\text{wire}_{\mu \neq \nu}} dx \, P(x, x),
\]

where \( \phi_{\mu\nu} = a_d \xi_{\mu\nu} \) is the localisation length in the infinitely long wire. From now, we will drop the primes on the vertices connected to the reservoirs, as there will be no possible confusion below.

### 3.2. Nonlocality leading to positive WL correction

An interesting consequence of the nature of the weighting factors was pointed out in Ref. [61]: since the weight \( \Delta G^{\text{class}}(\phi) \) may change in sign for certain wires, the WL correction to some transmission coefficient may become positive. Such an example is shown in Fig. 4, which has

\[
\Delta T_{12} \approx \frac{1}{3} \left( 1 + N_x \frac{L}{\xi_{12}} \right).
\]

valid for \( N_x \) long wires (\( \xi_{xx}, l_x \)). This WL correction may become positive for sufficiently large \( N_x \), as a striking illustration of the nonlocality of quantum transport.

### 3.3. Effect of nonlocality on the four-terminal resistances

We have demonstrated that a relation similar to (26) holds for the 4TRs, Eq. (9) (cf. Appendix B). We illustrate this formula by considering the resistances \( R_{12,12} \) and \( R_{12,34} \) characterizing the network of Fig. 5.

A remarkable consequence of quantum nonlocality is the possibility of large WL correction to the conductance, induced by the presence of long 1D contacts [52]. To illustrate this idea, we analyse the WL correction to the two resistances \( R_{12,12} \) and \( R_{12,34} \) for

\[
\frac{\Delta R_{12,12}}{R_{12,12}} = \frac{2}{L^2} \left( \int_{(a)} + \int_{(b)} + \int_{(c)} \right) dx \, P(x, x),
\]

where \( L = l_a + l_b + l_c \). The explicit expression of the integral over a wire in terms of the network properties can be found in Ref. [61]. Below, we only analyse limiting behaviours.

**Weakly coherent limit** \( L_c \ll l_a, l_b, l_c \): We get in this case

\[
\frac{\Delta R_{12,12}}{R_{12,12}} \approx \frac{L_c}{L} - \frac{2}{3} \left( \frac{L_c}{L} \right)^2,
\]

up to exponentially small corrections. If we compare this expression with (28), we remark that the presence of the two long wires \( (d) \) and \( (f) \) only affects the subleading term of the WL correction. The difference correction is explained by the depletion of the Cooperon at the vertices, which behaves as

\[ P(x, x) = L_c \left( 2 - \frac{L_c}{L} \right) e^{-2L_c|x|} \]

at distance \( x \) of the vertex. The correction to the bulk result is

\[ -4 \times (2L_c)^2 \int_0^\infty dx L_c e^{-2L_c|x|} = - (2/3)(L_c/L)^2. \]

**Fully coherent limit** \( L_c \to \infty \): We only discuss two limiting cases as the general expression is rather cumbersome. When the
two wires \(d\) and \(f\) are very long, \(l_0, l_f \ll l_a, l_b, l_c\), they play no role and we recover the universal result for the two-terminal wire \([\text{limit } L_0 \rightarrow \infty]\) of Eq. (28): \(\Delta R_{12,12}/(\xi_{12,12}^2)^2 \approx 1/3\).

In the opposite limit \(l_a, l_0, l_f \ll l_c, l_b, l_c\), the phase coherence is broken at the level of the vertices due to the vicinity of the large contacts. We get the result \(\Delta R_{12,12}/(\xi_{12,12}^2)^2 = (1/3)l_0^2 + l_f^2 + l_c^2)/(l_a + l_f + l_c)^2\), which corresponds to the addition of resistances \(\Delta R_{12,12} = \Delta R_a + \Delta R_f + \Delta R_c\) for three independent wires with \(\Delta R_i = (1/3)(l_i/\xi_{loc})^2\), etc.

Varying the length of the wires \(d\) and \(f\) hence allows us to cross over between the fully quantum regime where the wires \((a) + (b) + (c)\) should be considered as a whole and the regime where the three wires are independent and their resistances may be added according to the classical Kirchhoff law. This is an illustration of the idea introduced by Büttiker to describe dephasing in a fully coherent system by introducing fictitious voltage probes [15].

3.3.3. Four-terminal resistance \(R_{12,34}\): nonlocality leading to large WL correction

Classical resistance reads \(R_{12,34}^{\text{class}} = l_3/\xi_{loc}\) which immediately shows that the WL correction is given by an integral of the Cooperon over the wire \((b)\) only:

\[
\Delta R_{12,34}^{\text{class}} \\
\frac{(\xi_{12,34}^2)^2}{(R_{12,34})^2} = 2 \int_{\text{wire } (b)} dx \, p(x, x),
\]

(31)

Incoherent connecting wires \(l_a \ll l_c, l_b, l_f\): Using again the expression of the integral of the Cooperon integrated in a wire \([61]\), we obtain the explicit expression

\[
\Delta R_{12,34}^{\text{class}} \\
\frac{(\xi_{12,34}^2)^2}{(R_{12,34})^2} = -1 + \sqrt{r_b} \coth \sqrt{r_b} \biggl[ \frac{1}{r_b^2} + \frac{2}{4 \coth \sqrt{r_b} + 5} \left[ 1 - \sqrt{r_b} \coth \sqrt{r_b} \left( \coth \sqrt{r_b} + 2 \left( \coth \sqrt{r_b} - \frac{1}{\sinh \sqrt{r_b}} \right) \right) \right] \biggr],
\]

(32)

where \(r = l_0^2/\xi_{loc}^2\). The first term is the result for an isolated wire of length \(l_0\), Eq. (28). The second term originates from the nonvanishing value of the Cooperon at the two vertices, i.e. we can interpret this term as coming from the modification of the boundary conditions for the wire \((b)\) induced by the presence of the connecting wires \((a), (c), (d)\) and \((f)\). After a little bit of algebra, we obtain\(^4\)

\[
\Delta R_{12,34}^{\text{class}} \\
\frac{(\xi_{12,34}^2)^2}{(R_{12,34})^2} = \frac{l_0}{l_b} \frac{5 \coth(l_0/l_b) + 4 - 3 l_0/l_b}{4 \coth(l_0/l_b) + 5}.
\]

(33)

\(\bullet\) For a long wire \(l_0 \ll l_b\) we get the small correction

\[
\Delta R_{12,34}^{\text{class}} \\
\frac{(\xi_{12,34}^2)^2}{(R_{12,34})^2} = \frac{l_0}{l_b} \frac{1}{3} \left( \frac{l_0}{l_b} \right)^2,
\]

(34)

up to exponentially small corrections. The dominant term \(l_0/l_b \ll 1\), coincides with the two-terminal measurement for the wire of length \(l_0\), Eq. (28). As for \(\Delta R_{12,12}\), the presence of the connecting wires itself only through the coefficient of the subleading term \((l_0/l_b)^2\), cf. (28) or (30). The subleading correction is half of the one obtained for \(\Delta R_{12,12}\), Eq. (30), as only the depletion of the Cooperon at the two boundaries of the wire

\[^4\] This result was obtained by Santhanam [52], although this paper does not provide a detailed discussion on how the Cooperon must be integrated in a complex geometry.

Fig. 6. A wire of length \(l_b\) connected to 2N reservoirs by long wires. WL correction \(\Delta R_{12,34}/(\xi_{12,34}^2)^2\) is controlled by electronic trajectories starting from wire \((b)\) which may explore the wires over long distances compared to \(l_b\).

(b) contributes.

\(\bullet\) For a short (coherent) wire \(l_b \ll l_c\), quite remarkably, we obtain a large WL correction to the four-terminal conductance:

\[
\Delta R_{12,34} \\
\frac{(\xi_{12,34}^2)^2}{(R_{12,34})^2} \approx \frac{l_c}{2 l_b} \gg 1.
\]

(35)

In this case the presence of the connecting wires strongly affects the Cooperon inside the wire, which is the reason for the large WL correction. In a wire connected to two large reservoirs, the size of the electronic trajectories contributing to the WL is bounded by the length of the wire, which leads to a saturation of \(\Delta g\) as \(l_c \rightarrow \infty\). On the contrary, in the four-terminal configuration, the electronic trajectories can explore the connecting wires \((a), (c), (d)\) and \((f)\) on large scales compared to \(l_b\), which is the physical origin for the large WL.

Eq. (35) characterises the WL correction to the conductance, in a four probe configuration. Although the correction to the conductance may be large, we remark that the relative correction

\[
\Delta R_{12,34} \\
\frac{(\xi_{12,34}^2)^2}{(R_{12,34})^2} \approx \frac{l_c}{2 l_b} \gg 1
\]

(36)

is always small, as the validity of the perturbative treatment requires \(l_c \ll \xi_{loc}\).

In order to better understand the difference between the two results (34) and (35), it is instructive to consider the multiterminal network of Fig. 6. The WL correction to \(R_{12,34}\) is given by (31) as well. For the calculation, the key point is that, at a vertex \(x\) from which issue \(m_a\) long wires (longer than \(L_a\)), the value of the Cooperon is \(p(x, x) \approx L_a/m_a\) (cf. Ref. [59] and Appendix C). In the weakly coherent wire limit \((l_b \ll l_c\), the Cooperon inside the wire \((b)\) is \(p(x, x) \approx L_b/2\), except at a distance \(\lesssim l_c\) from the vertices. Integration of the Cooperon leads to \(\Delta R_{12,34}/(\xi_{12,34}^2)^2 \approx l_c/l_b \ll 1\), similar to Eq. (34).

In the coherent limit \((l_{a, b, f} \ll l_c\)\), with connecting wires still longer than \(l_{a, f}\), the Cooperon is almost uniform inside the wire, with a value \(p(x, x) \approx L_b/(2N_0)\), where \(2N_0\) is the effective coordination number. We get in this case the large WL correction \(\Delta R_{12,34}/(\xi_{12,34}^2)^2 \approx L_b/(N_0 l_b)\), which reduces to (35) for \(N_0 = 2\). This argument provides the interpretation of the factor 1/2 in Eq. (35).

Fully coherent limit \(l_c \rightarrow \infty\): Coming back to the simple network of Fig. 5, it is also interesting to consider the fully coherent limit. Using again the expression of the integral \(f_{\text{wire}} dx \, p(x, x)\) given in \([61]\), some algebra leads to

\[
\Delta R_{12,34}^{\text{class}} \\
\frac{(\xi_{12,34}^2)^2}{(R_{12,34})^2} = \frac{1}{3} + \frac{2}{3} l_{b, f} + l_{a, f} + 3 l_{a, f} l_{b, f}/l_b - l_{a, f} l_{c, f}/l_b - l_{b, f} l_{c, f} + l_{c, f} l_{c, b}.
\]

(37)

where \(1/l_{a, f} = 1/l_b + 1/l_{c, f}\), etc. For short connecting wire \(l_{a, f} \ll l_b\), \(l_a, l_g, l_f \ll l_b\) we recover the well known universal result \(\Delta R_{12,34}/(\xi_{12,34}^2)^2 = 1/3\) corresponding to a coherent wire between
two large contacts, as expected. In the other limit, $l_b \ll L = l_c = l_f$, we obtain a large correction:
\[ \Delta R_{\text{class}} \approx \frac{1}{2} \frac{l_b}{L} \gg 1 \]  
(38)

similar to (35) in which $L \rightarrow l_b$ (i.e. the cutoff limiting the trajectories exploring the connecting wires is not the phase coherence length but the distance $l_b$ to the reservoirs).

The possibility for large WL correction was pointed out in Ref. [31] in the rather academic situation of an isolated wire.\(^4\) A more precise discussion was provided by Santhanam [52] for the case we have considered here. As we already mentioned, this has the same origin as the large voltage fluctuations due to long coherent excursions of charge carriers in the voltage probes emphasised by Büttiker [16] (and also in Ref. [13]). Although the observation of large WL corrections has not been reported so far, to the best of our knowledge, large resistance fluctuations have been observed in several experiments [13,56], with the same physical origin.

4. Fluctuations and correlations

Mesoscopic (interference) phenomenon is more pronounced when the system size is reduced down to a size comparable to the phase coherence length $L$; the quantum contribution to the dimensionless conductance $\Delta g = g - g_{\text{class}}$ of a fully coherent conductor presents fluctuations $\delta g = g - \langle g \rangle \sim 1$ of the same order than the average $\langle g \rangle \sim 1$ (the WL). For this reason, the characterisation of conductance fluctuations/correlations has attracted considerable attention both experimentally [69,70,74,72] and theoretically [9,47,75] (for reviews, see [57,73,2]). The experiments are usually performed in the four-terminal configuration, which has brought the question of the role of the voltage probes [13,56]. Devices similar to the wire of Fig. 5 were studied in these experiments. The nonlocal nature of quantum transport is particularly striking by considering the symmetric and antisymmetric parts of the resistance:
\[ R_S = \frac{1}{2} \left( R_{12,34} + R_{34,12} \right) \]  
(39)
\[ R_A = \frac{1}{2} \left( R_{12,34} - R_{34,12} \right), \]  
(40)

which were shown to present different behaviours as a function of the ratio $l_b/L$ (symmetrisation is done with respect to exchange of current and voltage probes or, thanks to Eq. (10), to magnetic field reversal). Whereas $R_A$ is a relatively flat function of $l_b/L$, the symmetric resistance presents a clear crossover at $L \sim l_b$ see Fig. 8 (note that in a weakly disordered metal with a small enough magnetic field, we can ignore the classical magnetoresistance caused by the Lorentz force. As a consequence $R_{\text{class}}^A = 0$).

The study of nonlocality of voltage fluctuations and/or transmission probabilities in multiterminal devices was considered theoretically by Maekawa et al. [49] and Büttiker [16] (for a three terminal device) by different approaches. The importance of long range potential correlations was later emphasised by Kane et al. [44], which has led to reconsider the study of voltage fluctuations in an illuminating paper of Kane et al. [43], and also in Refs. [38,39] (note also the numerical study [11]). Finally we point out few theoretical works on the ring configuration [42,28] relevant for the experiments aforementioned, and specifically studied in Ref. [46].

In the following we derive formulae analogous to (8) and (9) for the correlations of transmissions and the correlations of four-terminal resistances:
\[ \langle \Delta R_{\alpha\beta} \Delta R_{\gamma\delta} \rangle \]  
(41)

We will apply our results to the analysis of the resistances $R_S$ and $R_A$ characterizing the four-terminal wire of Fig. 5. We will show that our formalism allows us to recover the results of Refs. [33,48,39] straightforwardly.

4.1. Conductance correlations

Expression of the conductivity correlations in simple geometries can be found at several places [9,2]. In networks, the correlations of transmission coefficients $\langle \delta T_{\alpha\beta}(\mathcal{B}) \delta T_{\gamma\delta}(\mathcal{B}') \rangle$ (i.e. conductance matrix elements) are given by four contributions [65]:
\[ \langle T_{\alpha\beta}(\mathcal{B}) \rangle = \frac{4}{\pi \epsilon^2} \int \frac{d\omega \delta_1(\omega)}{\omega} \int_{\text{Network}} \frac{dx \, dx'}{dx} \frac{\partial P_{\alpha\beta}(x, x')}{\partial \omega} \frac{\partial P_{\alpha\beta}(x', x)}{\partial \omega} \]  
(42)
\[ \langle T_{\alpha\beta}(\mathcal{B}) \rangle^\prime = \frac{4}{\pi \epsilon^2} \int \frac{d\omega \delta_1(\omega)}{\omega} \int_{\text{Network}} \frac{dx \, dx'}{dx} \frac{\partial P_{\alpha\beta}(x, x')}{\partial \omega} \frac{\partial P_{\alpha\beta}(x', x)}{\partial \omega} \]  
(43)
\[ \langle T_{\alpha\beta}(\mathcal{B}) \rangle^\prime^\prime = \frac{2}{\pi \epsilon^2} \int \frac{d\omega \delta_1(\omega)}{\omega} \int_{\text{Network}} \frac{dx \, dx'}{dx} \frac{\partial P_{\alpha\beta}(x, x')}{\partial \omega} \frac{\partial P_{\alpha\beta}(x', x)}{\partial \omega} \]  
(44)
\[ \langle T_{\alpha\beta}(\mathcal{B}) \rangle^\prime^\prime^\prime^\prime = \text{same as } \langle T_{\alpha\beta}(\mathcal{B}) \rangle^\prime^\prime, \]  
(45)

where we used the same notation as before, $P_{\alpha\beta}(x, x')$, in order to designate the Diffuson measured at a distance $\epsilon_{\alpha}$ of the vertex $\alpha$. The function $\delta_1(\omega)$ is a normalised function\(^6\) of width $\Delta \omega \sim T$ with $\delta_1(0) = 1/(6T)$. Several remarks:

- As for the WL, note that these expressions are of great generality, and not only valid for networks of quasi-1D wires; they only assume contacts of quasi-1D nature. For a more general situation, one has to replace the derivatives as $\partial P_{\alpha\beta}(g, x) \frac{\partial P_{\alpha\beta}(g, x)}{\partial \omega} \rightarrow \overline{V} P_{\alpha\beta}(g, r) \overline{V} P_{\alpha\beta}(g, r',)$.

- The Diffuson $P_{\alpha\beta}(x, x')$ connecting the contacts to the bulk, and providing the weights to attribute to each wire, obeys the classical diffusion equation (13). The Cooperon and Diffuson $P_{\alpha\beta}(x, x')$, which describe phase coherent properties, are solutions of the diffusion equation:
\[ \frac{1}{L_x} \int_{-l_x}^{l_x} - \frac{\partial^2}{\partial x^2} \int_{-l_x}^{l_x} p_{\alpha\beta}(x, x') = \delta(x - x') \]  
(46)

where the covariant derivative $D_x = \partial_x - 2 i e A_x(x)$ involves the vector potential $A_x = |A + A'|/2$ for Diffuson ($A_\alpha$) and Cooperon ($A_\alpha$).

- The penetration of the magnetic field in the wire is taken into account through the substitution $1/L_x^2 \rightarrow 1/L_x^2 + 1/(\omega^2 B^2)/2$, as for the WL (Section 3.1).

- These expressions are based on the current conserving

\[^{6}\text{Its precise form is } \delta_1(\omega) = F(\omega/2T)/2T \text{ with } F(x) = x \coth x - 1/\sinh x.\]
expressions for the conductivity correlations, given by the procedure of Kane et al. (i.e. we combine results of Refs. [9] and [44]).

- Transmission correlations are related to conductivity tensor correlations, $\sigma_{\alpha \beta} = e^2 \rho_{\alpha \beta}$. The two first contributions (42) and (43) which correlate the indices of the two transmissions are interpreted as diffusion constant correlations $(e^2 \rho_{\alpha \beta})^2 \delta \rho_{\alpha \beta} \delta \rho_{\alpha \beta}$, while the two last contributions (44) and (45), which do not correlate indices, are related to density of states fluctuations $(e^2 \rho_{\alpha \beta})^2 \delta \alpha \delta \beta$.

- The products of Diffusions may be related to derivative of classical transport coefficients, as it was done for the WL. For example

$$\frac{1}{e^2} \frac{\partial P_{\alpha}(x, t) \partial P_{\beta}(x, t)}{\partial x} = \frac{1}{\xi_{\text{loc}}} \frac{\partial T_{\alpha \beta}}{\partial x}$$

when $x$ belongs to the wire $i$, etc.

### 4.2. Four-terminal resistance correlations

We now simplify the above expressions by neglecting the effect of thermal smearing, $\delta T(\omega) \rightarrow \delta(\omega)$. Thermal effect will be described later in Section 4.3.3. We deduce the correlations for the 4TRs (cf. B.4):

\[
\langle R_{\alpha \beta \mu \nu}, R_{\alpha \beta', \mu' \nu'} \rangle \equiv \frac{4}{\xi_{\text{loc}}} \sum_{i j} \left[ \frac{\partial \rho_{\alpha \beta}}{\partial l_i} \frac{\partial \rho_{\alpha \beta}}{\partial l_j} \right] \int_{(i)} dx \int_{(v)} P_{\alpha}(x, x')^2
\]

(48)

\[
\langle R_{\alpha \beta \mu \nu}, R_{\alpha \beta', \mu' \nu'} \rangle \equiv \frac{2}{\xi_{\text{loc}}} \sum_{i j} \left[ \frac{\partial \rho_{\alpha \beta, \mu \nu}}{\partial l_i} \frac{\partial \rho_{\alpha \beta, \mu \nu}}{\partial l_j} \right] \int_{(i)} dx \int_{(v)} P_{\alpha}(x, x')^2
\]

(49)

\[
\langle R_{\alpha \beta \mu \nu}, R_{\alpha \beta', \mu' \nu'} \rangle \equiv \text{same as } \langle \cdots \cup \rangle \text{ with } P_{\alpha} \rightarrow P_{\alpha, \beta},
\]

(50)

where $\xi_{\text{loc}} = a N_c \ell_p$ is the localisation length for the infinitely long wire. As there will be no possible confusion, we now adopt the simpler notation $P_{\alpha} = P_{\alpha, \beta, \mu, \nu}$. Note that, as for the transmission correlations, only the contributions $\langle \cdots \cup \rangle$ and $\langle \cdots \cup \rangle$ correlate the indices in a non-trivial way. As for the WL, the contributions of each wires are weighted by classical quantities.

### 4.3. Four-terminal resistances in a multiterminal wire

We apply our formalism to the analysis of the 4TR correlations in a wire connected to several voltage probes, like the one represented in Fig. 5. Let us first recall the expressions of the classical resistances that will be needed to compute the weights in Eqs. (48)–(51):

$$R_{34,34}^{\text{class}} = (l_3 + l_4 + l_f + l_f) / \xi_{\text{loc}}$$

(52)

$$R_{12,34}^{\text{class}} = (l_1 + l_2) / \xi_{\text{loc}}$$

(53)

where $\xi_{\text{loc}}$ is the localisation length in the infinitely long wire. As they were considered in Ref. [13], we will study the symmetric and antisymmetric resistances $R_{\alpha \mu}$, which exhibit remarkable behaviours. The relations between correlators are

$$\langle \delta R_{34,34}^{\text{class}} \rangle = \frac{1}{2} \left( \langle \delta R_{12,12}^{\text{class}} \rangle \pm \langle \delta R_{34,34}^{\text{class}} \rangle \right).$$

(55)

We will first consider the limit of strong magnetic field ($L_{\phi} \ll L_{\phi}$), when the Cooperon contributions (49) and (51) are suppressed. The effect of a small field will be discussed in Section 4.3.4.

#### 4.3.1. Weakly coherent regime $L_{\phi} \ll L_{\phi}$

**Fluctuations** $\langle \delta R_{34,34}^{\text{class}} \rangle$: The contribution (48) is explicitly

$$C_{1,1} = \frac{4}{\xi_{\text{loc}}} \sum_{i j} \left[ \frac{\partial \rho_{\alpha \beta}}{\partial l_i} \frac{\partial \rho_{\alpha \beta}}{\partial l_j} \right] \int_{(i)} dx \int_{(j)} dx \langle \rho_{\alpha}(x, x')^2 \rangle$$

(54)

The wire weights are all equal to $1 / \xi_{\text{loc}}^2$ and imply $x \in (a, b, c)$ and $x' \in (d, e, f)$, thus this contribution can be splitted into nine terms $C_1 = C_{1,1} + \cdots + C_{1,8}$ of four different types (57), (59), (62), and (63). The first term involves a double integral in the central wire:

$$C_{1,1} \rightarrow a \begin{array}{c} x \ \ x' \ \ c \\ d \ \ b \ \ f \end{array}$$

(57)

$$C_{1,1} = \frac{4}{\xi_{\text{loc}}} \int_{(b)} dx \int_{(b)} dx \langle \rho_{\alpha}(x, x')^2 \rangle$$

$$\approx \frac{4}{\xi_{\text{loc}}} l_b \int dx (x - x') \langle \rho_{\alpha}(x, x')^2 \rangle \approx \frac{L_{\phi}^2}{\xi_{\text{loc}}},$$

(58)

where we have used the expression $\langle \rho_{\alpha}(x, x') \rangle \approx (L_{\phi}/2) \exp \left[ -(x - x') / L_{\phi} \right]$ valid in the bulk (i.e. the expression obtained in an infinitely long wire), as the presence of the wires $(a, c, f)$ only affect the diffusion at distance $\ll L_{\phi}$ from the two vertices.

Then four terms, $C_{1,2}$ to $C_{1,5}$, involve integration in neighbouring wires with one coordinate in the wire $(b)$ and the other in a connecting wire, like

$$C_{1,2} \rightarrow a \begin{array}{c} x \ \ x' \ \ c \\ d \ \ b \ \ f \end{array}$$

(59)

$$C_{1,2} = \frac{4}{\xi_{\text{loc}}} \int_{0}^{a} dx \int_{0}^{a} dx \langle \frac{l_{b}}{3} e^{-\sqrt{\pi} x^2 (x') / l_{b}^2} \rangle^2$$

(60)

leading to

$$C_{1,2} = \cdots = C_{1,5} \approx \frac{1}{9} \left( \frac{l_{b}}{\xi_{\text{loc}}} \right)^4.$$  

(61)

Two terms involve integration in neighbouring long connecting

---

7 The equality $C_{1,2} = \cdots = C_{1,5}$ holds for connecting wires much longer than $L_{\phi}.$
wires, like
\[ C_{1,6} \rightarrow x \]
\[ a \quad d \quad b \quad x' \quad \quad \quad \quad \quad f \quad e \]
(62)

When \( L_v \ll l_p \), we have \( C_{1,6} = C_{1,7} \approx C_{1,2} \). The two last contributions are of the kind
\[ C_{1,8} \rightarrow x \]
\[ a \quad d \quad b \quad f \quad x' \quad \quad \quad \quad \quad c \]
(63)

and are exponentially suppressed:
\[ C_{1,8} = C_{1,9} \sim \frac{L_v}{\xi_{\text{loc}}} e^{-2b/l_v}. \]
(64)

The contribution \( \langle \delta R_{12,34}^2 \rangle^{(3)} \) is simpler to discuss as it involves the weight
\[ \frac{\delta R_{12,34}^{\text{class}}}{\delta l_1} \frac{\delta R_{12,34}^{\text{class}}}{\delta l_2} \]
leading to the constraint \( x, x' \in (b) \). Thus we immediately get
\[ \langle \delta R_{12,34}^2 \rangle^{(3)} \approx \frac{L_v}{\xi_{\text{loc}}} \frac{L_v}{\xi_{\text{loc}}} \ll 1. \]
(65)

Correlations \( \langle \delta R_{12,34} \delta R_{34,12} \rangle \): The weights of the contribution \( \langle \cdot \rangle^{(1)} \) are
\[ \frac{\delta R_{12,34}^{\text{class}}}{\delta l_1} \frac{\delta R_{12,34}^{\text{class}}}{\delta l_2} \]
which requires once again \( x, x' \in (b) \). Therefore
\[ \langle \delta R_{12,34} \delta R_{34,12} \rangle^{(1)} = C_{1,1}. \]

The contribution \( \langle \cdot \rangle^{(3)} \) takes the same form, hence
\[ \langle \delta R_{12,34} \delta R_{34,12} \rangle^{(3)} = \frac{1}{2} C_{1,1}. \]

The correlations are given by summing these two terms
\[ \langle \delta R_{12,34} \delta R_{34,12} \rangle = \frac{3}{2} C_{1,1} \], which coincides exactly with the dominant term of the fluctuations.

Conclusion: If we now gather all these results we obtain
\[ \langle \delta R_{3}^2 \rangle = \frac{3}{2} C_{1,1} + C_{1,2} + \ldots + C_{1,9} \ll \frac{3}{2} \frac{L_v^2}{\xi_{\text{loc}}^2} \]
(66)

\[ \langle \delta R_{3}^2 \rangle = \frac{C_{1,2} + \ldots + C_{1,9}}{2} \approx \frac{1}{3} \frac{L_v}{\xi_{\text{loc}}} \]
(67)

\( \delta R_3 \) grows with \( l_p \) whereas \( \delta R_3 \) is independent of the distance between the two voltage probes, in agreement with the experiment (see Fig. 8).

4.3.2. Coherent limit \( l_p \gg L_v \)

We now discuss the case where the central wire is coherent, with long connecting wires \( (l_p, l_1, l_2, l_f \ll L_v) \). We can write that the Diffuson at a vertex of coordination number 4. Thus \( R_3(x, x') \approx L_v/4 \) for \( x, x' \in (b) \). It decays exponentially over the distance \( L_v \) in the connecting wires: \( R_3(x, x') \approx (L_v/4)\exp(-|x-x'|/L_v) \) for \( x \in (a) \) and \( x' \in (c) \), etc. (cf. Appendix C). As a result we see that the fluctuations are dominated by four terms, \( C_{1,6} \ldots C_{1,9} \) corresponding to cases like (62) or (63), when \( x \) and \( x' \) are both integrated over the long distance \( l_p \) in the connecting wires: \( \langle \delta R_{12,34}^2 \rangle = (1/4)L_v/\xi_{\text{loc}}^2 \), i.e. the four-terminal conductance fluctuations are large:
\[ \langle \delta R_{12,34}^2 \rangle \simeq \frac{1}{4} \left( \frac{L_v}{\xi_{\text{loc}}} \right)^4 \gg 1. \]
(68)

The correlations are much smaller as they involve integrations of \( x \) and \( x' \) in the central wire only \( \langle \delta R_{12,34} \delta R_{34,12} \rangle \approx (3/8)L_v/\xi_{\text{loc}}^2 \), i.e.
\[ \langle \delta R_{12,34} \delta R_{34,12} \rangle \simeq \frac{3}{8} \left( \frac{L_v}{\xi_{\text{loc}}} \right)^2 \gg 1. \]
(69)

Finally we get
\[ \langle \delta R_3^2 \rangle \approx \langle \delta R_3^2 \rangle \approx \frac{1}{8} \left( \frac{L_v}{\xi_{\text{loc}}} \right)^4. \]
(70)

Fluctuations are independent of the distance \( l_p \) between the voltage probes.

4.3.3. Thermal smearing

In the two previous paragraphs, we have reproduced the main conclusions of Herschfield [38] by simpler arguments based on the analysis of the wire weights. Although the qualitative change of behaviour between \( \delta R_3 \) and \( \delta R_3 \) at \( l_p \sim l_p \) agrees with the experiment of Benoit et al. [13] (Fig. 8), it was noticed that thermal smearing, not accounted for by Hershfield, is important in the experiment. This corresponds to the case where the thermal length \( l_T = \sqrt{D/T} \) is smaller than \( l_p \).

There is also a fundamental reason to consider this regime: in the low temperature regime \( T \lesssim 1 \) K and in the absence of external degrees of freedom like magnetic impurities, the decoherence is dominated by electronic interactions. As a consequence one has \( l_p \ll l_p \) (see [7,6,2,5,6,9] and references therein).

In this regime the contribution \( \langle \cdot \rangle^{(3)} \) is negligible compared to \( \langle \cdot \rangle^{(1)} \).

**Weakly coherent limit \( l_p \ll l_p \):** We should repeat the analysis of the previous subsection by adding an integration over the frequency with the thermal function. The first contribution to the resistance fluctuations reads
\[ C_{1,1} = \frac{4}{\xi_{\text{loc}}^2} \int_0^\infty \tilde{T}(\omega) \times \int_0^\infty dx \int_0^\infty dx' \int_0^\infty dx' \int_0^\infty dx' \int_0^\infty dx' \int_0^\infty dx' \int_0^\infty dx'. \]
(71)

At finite temperature, the Diffuson involves exponentials like \( \exp(-\sqrt{y}x) \) with \( \sqrt{y} = 1/2 - \omega_0/\omega_0 \) where \( \omega_0/\omega_0 \leq 1/2 \). When \( \min(\omega_1, \omega_2) \ll \omega_0 \), the Diffuson decays rapidly inside the wire. \( \omega_0 \) is dominated by the integral in the bulk of the wire, which leads to a similar calculation as for an infinite wire. When \( L_p \ll l_p \) (the case \( L_p \ll l_p \) is similar to \( L_p = \infty \), and was discussed above), the thermal function can be considered as a broad function and simply replaced by \( \delta f(x) = 1/6T \). Using the expression of the Diffuson in bulk we get
\[ C_{1,1} \approx \frac{\pi}{3} \frac{L_v^2}{\xi_{\text{loc}}^2}. \]
(72)

This term dominates the fluctuations, therefore we recover the known expression [2] of the conductance fluctuations:
where $L = 0.15 \mu m$ ($450 \Omega$) and $5 \mu m$ ($12 \Omega$). Data from Ref. [56].

\[
\langle \delta R_{12,34}^2 \rangle_{\text{class}} \approx \frac{\pi L l_T^2}{3 l_b^4}.
\]

(76)

Next contributions are the six types of term (59) and (62). We can again argue that integration is dominated by $x$ and $x'$ close to the vertex and use the approximate given in Appendix C:

\[
C_{1,2} \approx \frac{4}{\pi l_{\text{loc}}} \int_{\delta}^{\pi} \frac{d\varphi}{\delta} \int_{0}^{\infty} dx \int_{0}^{\infty} dx' \frac{P_0^d(x, x')}{l_{\text{loc}}} \frac{P_0^{dd}(x, x')}{l_{\text{loc}}},
\]

(77)

where $P_0^d(x, x') \approx 1/(3\sqrt{2}) \exp(-\gamma/(2x))$ with $\gamma = 1/L_l^2 - \ln|D|$. Some algebra gives

\[
C_{1,2} \approx \cdots \approx C_{1,2} \approx \frac{2}{27} \frac{L l_T^4}{l_b^4}.
\]

(78)

Correlations are obtained by similar arguments

\[
\langle \delta R_{12,34}^2 \rangle_{\text{class}} = C_{1,2},
\]

which coincide with the dominant term of the fluctuations.

Going back to the quantity of interest, we should replace (69) by $\langle \delta R^2 \rangle = \langle \delta R_{1,1}^2 \rangle + \langle C_{1,2} \rangle + \cdots + C_{1,3}/2$ whereas (70) still holds, thus

\[
\langle \delta R^2 \rangle = \frac{\pi L l_T^2}{3 l_b^4}.
\]

(79)

\[
\langle \delta R^2 \rangle = \frac{2}{9} \frac{L l_T^4}{l_b^4}.
\]

(80)

We reproduce similar conclusions as for $L_T = \infty$, i.e. fluctuations $\delta R_{1,2}$ growing with the distance $l_{\text{loc}}$ between the voltage probes, while the fluctuations $\delta R_{1,3}$ are independent of $l_b$.

Coherent limit $l_b \ll l_\varphi \ll l_\delta$: As in the case $l_b \ll l_\varphi$ analysed previously, the fluctuations are dominated by four terms where $x$ and $x'$ are integrated over two long connecting wires. It is simply given by multiplying (78) by a factor $(3/4)^2$, which accounts for the fact that the Diffuson feels an effective coordination number 4 instead of 3, hence $C_{1,6} = \cdots \approx C_{1,9} \approx (1/24)L l_T^2 l_b^2 / l_{\text{loc}}^4$. Thus

\[
\langle \delta R_{12,34}^2 \rangle / \langle \delta R_{12,34}^2 \rangle_{\text{class}} \approx (1/6)(1/4)(1/24)(L l_T^2 l_b^2 / l_{\text{loc}}^4) \leq 1.
\]

(81)

The analysis of the correlations $\langle \delta R_{12,34} \delta R_{14,32} \rangle = C_{1,1}$ is as follows: we write $P_0^d(x, x') \approx 1/(4\sqrt{2})$ when $x \neq x' \in (b)$, therefore $C_{1,1} \approx (1/4)(L l_T^2 l_b^2 / l_{\text{loc}}^4) \int_{\pi}^{\pi} \exp(-\gamma/(2x)) dx$. The presence of the function $\exp(-\gamma/(2x))$ is needed in order to cut off the contribution of large frequencies. When $L_T \ll L_\varphi$, some algebra gives

\[
\langle \delta R_{12,34} \delta R_{14,32} \rangle / \langle \delta R_{12,34}^2 \rangle_{\text{class}} \approx 1/6 \left( \frac{L l_T^2}{l_b^2} \right) \ln(l_\varphi/l_T).
\]

(82)

Quite surprisingly, we obtain a logarithmic dependence on $l_\varphi$ reminiscent of the 2D situation, although the system is 1D.

The symmetric and antisymmetric resistance fluctuations are

\[
\langle \delta R_S^2 \rangle \approx \langle \delta R_A^2 \rangle \approx \frac{1}{12} \frac{L l_T^2}{l_b^2}.
\]

(83)

4.3.4. Magnetic field dependence

Finally, we discuss the magnetic field dependence. The above results are valid when the magnetic field $B = B_0$ is larger than the correlation field, i.e. in a diffusive wire, when $B \ll B_0 \sim \phi_0 / l_\varphi$ where $w$ is the width of the wires and $\phi_0 = h/e$ the flux quantum. At small magnetic field, Cooperon contributions (49) and (51) must be taken into account as well. This can be done easily by noticing that, due to the wire weights, the role of fluctuations and correlations is exchanged for the two first contributions:

\[
\langle \delta R_{12,34}^2 \rangle_{(2)} = \langle \delta R_{12,34} \delta R_{14,32} \rangle_{(1)}
\]

\[
\langle \delta R_{12,34}^2 \rangle_{(2)} = \langle \delta R_{12,34}^2 \rangle_{(1)}
\]

while

\[
\langle \delta R_{12,34}^2 \rangle_{(4)} = \langle \delta R_{12,34} \delta R_{14,32} \rangle_{(2)}
\]

\[
\langle \delta R_{12,34}^2 \rangle_{(4)} = \langle \delta R_{12,34}^2 \rangle_{(3)}
\]

The substitution $P_2 \rightarrow P_1$ is simply achieved by doing $1/4 < 1/L_T^2 + 1/L_\varphi^2$, cf. Section 3.1 (the analysis of the correlations for different fields requires similar substitution in both the Diffuson and the Cooperon, cf. Section 4.1). We deduce straightforwardly the crossover functions. We only consider the weakly coherent regime, $l_\varphi \ll l_b$.

Regime $l_\varphi \ll L_T < l_b$: Using the previous calculations we obtain

\[
\langle \delta R_S^2 \rangle \approx \frac{3}{2} \frac{L l_T^2}{l_b^2} \left( 1 + \left[ 1 + (B/B_0)^2 \right]^{-1/2} \right)
\]

(84)

\[
\langle \delta R_A^2 \rangle \approx \frac{1}{3} \frac{L l_T^2}{l_b^2} \left( 1 - \left[ 1 + (B/B_0)^2 \right]^{-1} \right).
\]

(85)

where $B/B_0 = L_\varphi/L_T = (2\pi/\sqrt{3}) b w l_\varphi/\phi_0$. Whereas the symmetric resistance fluctuations are doubled at small field, the antisymmetric resistance vanishes as $\langle \delta R_A(B) \rangle = \langle \delta R_A(0) \rangle = \langle \delta R_A(B_0) \rangle = 0$.

Regime $L_T \ll l_b \ll l_\varphi$: In this case we obtain

\[
\langle \delta R_S^2 \rangle \approx \frac{\pi L l_T^2}{3 l_b^4} \left( 1 + \left[ 1 + (B/B_0)^2 \right]^{-1/2} \right)
\]

(86)

\[
\langle \delta R_A^2 \rangle \approx \frac{\pi L l_T^2}{3 l_b^4} \left( 1 - \left[ 1 + (B/B_0)^2 \right]^{-1} \right).
\]

(87)

Fig. 7. Mesoscopic (sample to sample) resistance fluctuations for multiconnected silicon inversion-layer narrow wire at $T = 400$ mK. The two samples have lengths $L = 0.15 \mu m$ ($R_{12,34} \approx 450 \Omega$) and $5 \mu m$ ($R_{12,34} \approx 12 \Omega$). Data from Ref. [56].

Fig. 8. Mesoscopic resistance fluctuations for multiconnected Au and Sb wires at $T = 40$ mK and 300 mK. Length $l$ varies from 0.2 pm to 4 pm. $l_\varphi$ is the distance between the voltage probes. Blue circles correspond to the symmetric resistance $\langle \delta R_S^2 \rangle_{(3)}$ and red diamonds to the antisymmetric resistance $\langle \delta R_A^2 \rangle_{(3)}$. The fluctuations for a wire of length $L_\varphi$ are denoted by $\delta R = (L_\varphi/l_\varphi)^2$. Data from Ref. [13].
(δR^2_φ/δβ)^2 \approx \frac{1}{12} \frac{L^2}{δν} \frac{β^2}{2β^2 + 2β^2}.

(87)

The antisymmetric resistance now vanishes at small field as δR(β) \sim (L_L L_l β^2 L_b W/φ_0) for β \rightarrow 0.

4.3.5. Experiments

We now discuss the experiments at the light of our results. In Fig. 7, we have reproduced the experimental data obtained by Skocpol et al. [56] for a Silicon inversion-layer narrow wire. We now denote by L the distance between the two voltage probes (denoted L_l above). The experimental result exhibits the behaviours obtained above: a growth of the fluctuations with the length, δR_{12,34} \sim L^{3/2} L^2 in the incoherent regime L_L \leq L, Eq. (67), and a saturation δR_{12,34} \sim (L_L/δυ)^2 in the coherent regime L_L \geq L, Eq. (71). Note that accounting for thermal broadening does not change this conclusion, cf. Eqs. (76) and (81).

Another remarkable experiment is the one of Benoît et al. [13], who analysed voltage fluctuations in Sb and Au narrow wires. Using the data given in this reference, we have plotted the fluctuations of the symmetric and antisymmetric resistances in Fig. 8. The most striking outcome is the comparison of the different behaviours for the symmetric and antisymmetric resistances.

Let us analyse the behaviours more into detail. We follow the line of Ref. [13], where thermal broadening was not taken into account. Voltage (sample to sample) fluctuations were measured and compared to the fluctuations obtained for a wire of length L_L. This corresponds to consider the ratio δR_{SFL}/δR_{FL} where

δR_{SFL} = (L_L/δυ)^2.

In the coherent limit (L_L \geq L), Eqs. (73) gives the value δR_{SFL}/δR_{FL} = 1/\sqrt{8} \approx 0.35 consistent with the experimental data.

In the incoherent limit (L_L \leq L), Eqs. (70) gives δR_{SFL}/δR_{FL} = 1/3\sqrt{2} \approx 0.577, whereas (69) leads to δR_{SFL}/δR_{FL} \approx \sqrt{3/2} L_L/δυ. Although the linear behaviour with \sqrt{L_L/δυ} agrees qualitatively with the experimental data, the prefactor obtained experimentally, ≈ 0.47, is significantly smaller than \sqrt{3/2} \approx 1.22.

The fact that the experimental values are smaller than the theoretical predictions for L_L \ll L \_F could be explained by the effect of thermal broadening, as the above calculation for L \_F \ll L \_L has led to an additional reduction factor \sqrt{L_L/L_L}. A more precise analysis would nevertheless be needed. Another aspect which could explain this quantitative disagreement concerns the determination of the phase coherence length, which was not obtained by a unique independent procedure in Ref. [13]. At the highest temperatures, the phase coherence length was obtained by independent WL measurements, which does not account for the fact that, in heavy metals like gold, when spin–orbit scattering is strong and magnetic impurities present, the phase coherence length involved in the weak localisation [40] and the conductance fluctuations [24] differ (see also [2]). The lack of a formula for the WL correction in the multiterminal wire did not allow the authors to extract L_L for the lowest temperatures. A more reliable procedure would have been to obtain L_L from another long wire (\gg L_L) made under the same conditions.

5. Conclusion

Since the pioneering work of Markus Büttiker, four-terminal resistance is now recognised as a quantity of major importance in mesoscopic physics (Ref. [14] is indeed Büttiker’s most cited article, with now almost 2000 citations). After having briefly reviewed several aspects of four-terminal resistances, we have focused the discussion on the analysis of quantum transport in networks of weakly disordered wires, for which explicit expressions for the quantum corrections to the conductances and the four-terminal resistances were discussed. We have seen that both the weak localisation correction and the correlations involve integrals of the Cooperon (and also of the Diffuson for the correlations) inside the wires, whose contributions must be weighted by derivatives of the classical coefficients: δR_{SFL}/δβ for the conductances and δR_{SFL}/δβ for the four-terminal resistances. Although we had discussed this earlier for the WL correction to the conductance [61], we have provided here a new interpretation of these coefficients in terms of “generalised conductances” G_{\alpha} relating the current in a wire i inside the network to the external potential V_i, precisely: δR_{SFL}/δβ \sim G_{\alpha} G_{\beta}.

We have illustrated the efficiency of our formalism by considering simple examples. Our determination of the four-terminal resistance correlations only involves simple calculations (which greatly simplify the analysis of Ref. [38] in particular). Moreover we have been able to consider the effect of thermal broadening, which was not studied so far. All the main dependences are summarised in Tables 1 and 2.

Table 1

<table>
<thead>
<tr>
<th>( L_\nu \approx \beta )</th>
<th>( L_\nu \approx L )</th>
</tr>
</thead>
<tbody>
<tr>
<td>fluct. _ corr.</td>
<td>fluct. _ corr.</td>
</tr>
<tr>
<td>L_\nu \approx L</td>
<td>(L_\nu/\beta)^2</td>
</tr>
<tr>
<td>L_\nu \approx L</td>
<td>(L_\nu/\beta)^2</td>
</tr>
<tr>
<td>L_\nu \approx L</td>
<td>(L_\nu/\beta)^2</td>
</tr>
</tbody>
</table>

Table 2

<table>
<thead>
<tr>
<th>( L_\nu \approx L )</th>
<th>( L_\nu \approx L )</th>
</tr>
</thead>
<tbody>
<tr>
<td>(\delta R^2_φ)</td>
<td>(\delta R^2_φ)</td>
</tr>
<tr>
<td>L_\nu \approx L</td>
<td>(L_\nu/\beta)^2</td>
</tr>
<tr>
<td>L_\nu \approx L</td>
<td>(L_\nu/\beta)^2</td>
</tr>
</tbody>
</table>

Appendix A. Conductance matrix, resistance matrix and four-terminal resistances

A.1. Conductance and resistance matrices

A natural way to characterise the linear response of a multiterminal structure is to introduce the conductance matrix relating the voltages at the contacts to the currents: L = \sum \delta G_{\alpha\beta} V_\beta. The matrix elements of the conductance matrix are not independent as they must satisfy two types of constraints: (i) current is conserved, thus \sum \delta G_{\alpha\beta} = 0 whatever the choice of external potentials. (ii) A
global shift of all voltages \( V_i \rightarrow V_i + U_i \) does not change the current. As a consequence \( \sum G_{\alpha\beta} = \sum G_{\beta\alpha} = 0 \). In other terms, introducing the vector \( X_{\alpha\beta}^T = (1, 1, \ldots, 1) \) where \( T \) denotes transposition, we rewrite the second condition as \( G_{00} = 0 \), i.e. \( G \) is a linear map acting in the vector space \( X_i = (\{ x \in \mathbb{R}^N | X_i \} | \mathbb{X} = 0) \). The current conservation rewrite \( X_{\alpha\beta}^T G = 0 \), meaning that \( G \) maps \( X_i \) onto itself. In the subspace \( X_i \), we may invert the conductance matrix, which leads to introduce the resistance matrix \( R \) relating the external currents to the voltages \( V = -RI \). The relation between the two matrices is thus \( RG = GR = X_{\alpha\beta}^T P_i \), where \( P_i = (1/N)X_{\alpha\beta}^T \) is the projector on the vector \( X_{\alpha\beta}^T \), where \( N \) is the number of contacts. We write more conveniently:

\[
\sum R_{\alpha\beta}G_{\beta\alpha} = \sum G_{\beta\alpha}R_{\beta\alpha} = \delta_{\alpha\beta} - \frac{1}{N}.
\]  
(A.1)

Since we will have to consider differences between potentials at various contacts, the relation

\[
\sum (R_{\alpha\beta} - R_{\beta\alpha})G_{\beta\alpha} = \sum G_{\beta\alpha}(R_{\alpha\beta} - R_{\beta\alpha}) = \delta_{\alpha\beta} - \delta_{\beta\alpha}
\]  
(A.2)

will be useful.

A.2. Four-terminal resistance

The four-terminal resistance provides information in the situation where current only flows through contacts \( \alpha \) and \( \beta \), \( I_\alpha = -I_\beta = I \), with all other currents vanishing, Eq. (5). The voltage at contact \( \mu \) is thus conveniently expressed in terms of the resistance matrix as \( V_\mu = (R_{\mu\alpha} - R_{\mu\beta})I \) and the voltage difference \( V_\nu - V_\mu = -R_{\mu\nu}I \) thus involves

\[
R_{\mu\nu} = R_{\mu\alpha} - R_{\mu\beta} - R_{\alpha\nu} + R_{\beta\nu}.
\]  
(A.3)

A useful relation is obtained by differentiating the relation (A.2) [43]:

\[
\delta R_{\mu\nu} = -\sum (R_{\mu\beta} - R_{\beta\mu})\delta G_{\beta\nu}(R_{\alpha\nu} - R_{\beta\nu}).
\]  
(A.4)

Appendix B. Averaging functions of the conductances

As we have discussed in the body of the text, the quantum contributions to the transport coefficients are naturally expressed for the conductance matrix, as linear response theory provides formulae for the conductivity or the conductance. On the other hand, resistances, which are in general complicated functions of the set of conductance matrix elements, are more easy to handle in most situations, and usually the relevant quantities in most experiment. We show in this appendix how one can go from the quantum contributions to the conductance matrix elements to the equivalent contributions to the four-terminal resistances.

Let us consider a general quantity, function of the conductance matrix elements \( A = f(G) \), where \( f(G) \) is a short notation for a function of all matrix elements \( f(G_{11}, G_{12}, G_{13}, \ldots) \) ; the quantity \( A \) may be for example a four-terminal resistance. We now show how the quantum contributions to the quantity \( A \) can be related to the quantum contributions to the conductance. For this purpose it is convenient to analyse the scaling with the number of channels \( N_c \), which is a large parameter.\(^9\) Considering the set of conductance matrix elements \( G_{\alpha\beta} \), the disorder average is given by two contributions: a classical term and the weak localisation correction: \( G_{\alpha\beta} = G^{\text{class}}_{\alpha\beta} + \Delta G_{\alpha\beta} \). Other quantum contributions of interest are the mesoscopic (sample to sample) fluctuations, defined by \( \Delta G_{\alpha\beta} = (G_{\alpha\beta} - G_{\alpha\beta}^{\text{class}}) \) and characterised by the correlation functions \( \langle \delta G_{\alpha\beta} \delta G_{\gamma\delta} \rangle \). As we have recalled in the text, the three quantities present the \( N_c \) dependence:

\[
G^{\text{class}}_{\alpha\beta} = O(N_c)
\]  
(B.1)

\[
\Delta G_{\alpha\beta} = O(N_c^0)
\]  
(B.2)

\[
\langle \delta G_{\alpha\beta} \delta G_{\gamma\delta} \rangle = O(N_c^0).
\]  
(B.3)

We now deduce two useful properties.

B.1. Property 1

Writing \( G_{\alpha\beta} = (G_{\alpha\beta}^{\text{class}} + \delta G_{\alpha\beta}) \), the fluctuation is smaller than the average, \( \langle \delta G_{\alpha\beta} \rangle = O(N_c^0) \), and vanishes on average by definition. As a consequence

\[
\langle f(G) \rangle = f((G)^{\text{class}}) + O(f \times N_c^{-2}).
\]  
(B.4)

B.2. Property 2

Using this property, we can split the average into a classical part and the WL correction: \( (G) = G^{\text{class}} + \Delta G \). Hence the average of the physical quantity takes the form

\[
\langle A \rangle = f(G^{\text{class}} + \Delta G) + O(f \times N_c^{-2})
\]

\[
= f(G^{\text{class}}) + \sum_{\mu,\nu} \frac{\partial f(G^{\text{class}})}{\partial G_{\mu\nu}} \Delta G_{\mu\nu} + O(f \times N_c^{-2}).
\]  
(B.5)

We identify the second term as the weak localisation correction to \( A \):

\[
\delta A = \sum_{\alpha,\beta} \frac{\partial A}{\partial G_{\alpha\beta}} \Delta G_{\alpha\beta}.
\]  
(B.6)

B.3. Property 3

If we consider two quantities \( A \) and \( B \) functions of the transmissions, following the same lines, we deduce the expression for the correlation function:

\[
\langle \delta A \delta B \rangle = \sum_{\alpha,\beta,\mu,\nu} \frac{\partial A}{\partial G_{\alpha\beta}} \frac{\partial B}{\partial G_{\mu\nu}} \langle \delta G_{\alpha\beta} \delta G_{\mu\nu} \rangle.
\]  
(B.7)

B.4. Applications

- WL correction to the four-terminal resistances: It is now straightforward to get the weak localisation correction to the four-terminal resistance. We deduce from the property (B.4)\(^7\)

\[
\sum_{\alpha,\beta} ((R_{\alpha\beta}^{\text{class}} - (R_{\alpha\beta}^{\text{class}}))(G_{\beta\mu}) = \delta_{\alpha\beta} - \delta_{\alpha\beta} + O(N_c^{-2})
\]

Then expanding the conductances as \( (G_{\alpha\beta}) = G^{\text{class}}_{\alpha\beta} + \Delta G_{\alpha\beta} \), and using (A.4) and the property

\[
\sum_{\beta} \frac{\partial A}{\partial G_{\beta\mu}} \frac{\partial A}{\partial G_{\alpha\nu}} \delta G_{\beta\mu} \delta G_{\alpha\nu}.
\]  
(B.7)
Using (26) and once again Eq. (A.4) we finally get (9). Note that this result could have been guessed from the heuristic argument presented in the introduction of Ref. [61].

**Correlations of four-terminal resistances:** The expressions for the correlation functions (48)–(51) are demonstrated by making use of the relation (A.4) with the property (B.7) (with the remark closing paragraph 4.1).

### Appendix C. Solution of the diffusion equation in the star graph

The solution of the diffusion equation

$$\frac{1}{a^2} \partial_x^2 P(x, x') = -4 \delta(x - x')$$

in the star graph with $m_i$ infinite wires is useful (Fig. 9). Details are given in Appendix D of Ref. [59]. The value of the Cooperon at the vertex $\alpha$ is inversely proportional to the coordination number of the vertex: $P_\alpha(a, a) = 1/m_{\alpha}^2$. When the two arguments belong to wires $x \in (i)$ and $x' \in (j)$, we get

$$P_\alpha(x, x') = \frac{1}{m_{\alpha}^2} e^{-\sqrt{\pi} xx'} + \frac{1}{\sqrt{\pi}} \sinh(\sqrt{\pi} xx') e^{-\sqrt{\pi} xx'}$$

where $x = \min(x, x')$, $x = \max(x, x')$ and the distance is measured at the vertex. At large distance of the vertex, we recover the result for the infinite wire $P(x, x') = \exp[-\sqrt{\pi} x - x']$.

### References


